

# FAN5307

## High-Efficiency Step-Down DC-DC Converter

### Features

- 95% Efficiency, Synchronous Operation
- Adjustable Output Voltage Option: 0.7V to 0.8V<sub>IN</sub>
- 2.5V to 5.5V Input Voltage Range
- Customized Fixed Output Voltage Options
- Up to 300mA Output Current
- Fixed Frequency 1MHz PWM Operation
- High-Efficiency, Power-Save Mode
- 100% Duty Cycle Low Dropout Operation
- Soft Start
- Dynamic Output Voltage Positioning
- 15µA Quiescent Current
- Excellent Load Transient Response
- 5-Lead SOT-23 Package
- 6-Lead MLP 3x3mm Package

### Applications

- Pocket PCs, PDAs
- Cell Phones
- Battery-Powered Portable Devices
- Digital Cameras
- Low Power DSP Supplies


### Description

The FAN5307, a high-efficiency, low-noise synchronous PWM current mode and Pulse Skip (Power-Save) mode DC-DC converter, is designed for battery-powered applications. It provides up to 300mA of output current over a wide input range from 2.5V to 5.5V. The output voltage can be either internally fixed or externally adjustable over a wide range of 0.7V to 0.8V<sub>IN</sub> by an external voltage divider. Custom output voltages are also available. Contact a Fairchild sales representative for customized output voltage options.

At moderate and light loads, pulse skipping modulation is used. Dynamic voltage positioning is applied, and the output voltage is shifted 0.8% above nominal value, for increased headroom during load transients. At higher loads, the system automatically switches to current mode PWM control, operating at 1 MHz. A current mode control loop with fast transient response ensures excellent line and load regulation. In Power-Save mode, the quiescent current is reduced to 15µA to achieve high efficiency and ensure long battery life. In shutdown mode, the supply current drops below 1µA.

The device is available in 5-lead SOT-23 and 6-lead MLP 3x3mm packages.

### Ordering Information

Part Number	Operating Temperature Range	V <sub>OUT</sub> (V)	Package	 Eco Status	Packing Method
FAN5307S18X	-40 to +85°C	1.8	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MP18X	-40 to +85°C	1.8	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel
FAN5307S15X	-40 to +85°C	1.5	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MP15X	-40 to +85°C	1.5	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel
FAN5307SX	-40 to +85°C	Adjustable	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MPX	-40 to +85°C	Adjustable	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel

 For Fairchild's definition of Eco Status, please visit: [http://www.fairchildsemi.com/company/green/rohs\\_green.html](http://www.fairchildsemi.com/company/green/rohs_green.html).

### Typical Applications

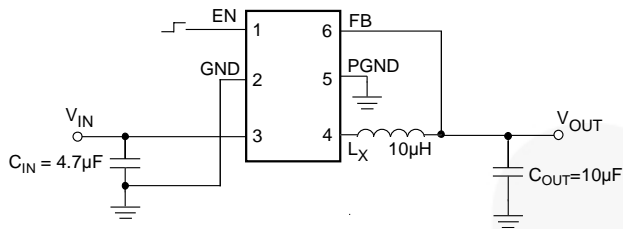


Figure 1. 6-Lead 3x3mm (MLP)

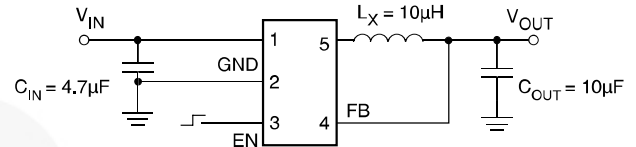


Figure 2. 5-Lead SOT-23

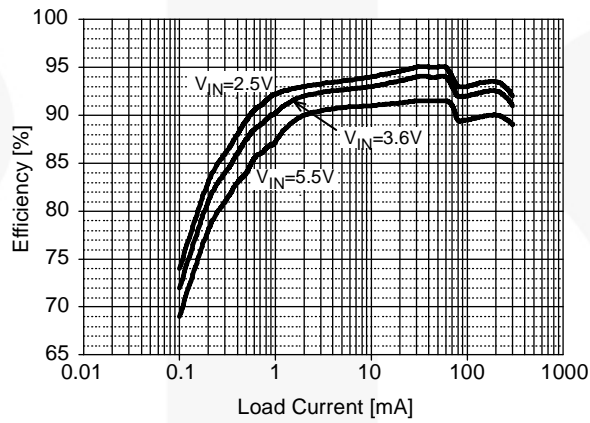


Figure 3. Efficiency vs. Load Current ( $V_{OUT}=1.8V$ )

## Pin Configuration

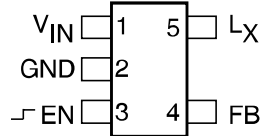


Figure 4. 5-Lead SOT-23

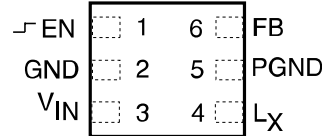


Figure 5. 6-Lead 3x3mm MLP

## Pin Definitions

### 5-Lead SOT-23

Pin #	Name	Description
1	V <sub>IN</sub>	<b>Supply Voltage Input.</b>
2	GND	<b>Ground.</b>
3	EN	<b>Enable Input.</b> Logic HIGH enables the chip; logic LOW disables the chip and reduces supply current to <math><1\mu\text{A}</math>. Do not float this pin. If the EN pin is tied to V <sub>IN</sub> , V <sub>IN</sub> must be ramped up faster than 5V/ms for V <sub>OUT</sub> to enter regulation.
4	FB	<b>Feedback Input.</b> In case of fixed-voltage options, connect this pin directly to the output. For an adjustable voltage option, connect this pin to the resistor divider.
5	L <sub>X</sub>	<b>Inductor Pin.</b> This pin is connected to the internal MOSFET switches.

### 6-Lead 3x3mm MLP

Pin #	Name	Description
1	EN	<b>Enable Input.</b> Logic HIGH enables the chip; logic LOW disables the chip and reduces supply current to <math><1\mu\text{A}</math>. Do not float this pin. If the EN pin is tied to V <sub>IN</sub> , V <sub>IN</sub> must be ramped up faster than 5V/ms for V <sub>OUT</sub> to enter regulation.
2	GND	<b>Reference Ground.</b>
3	V <sub>IN</sub>	<b>Supply Voltage Input.</b>
4	L <sub>X</sub>	<b>Inductor Pin.</b> This pin is connected to the internal MOSFET switches
5	PGND	<b>Power Ground.</b> The internal N-channel MOSFET is connected to this pin.
6	FB	<b>Feedback Input.</b> In case of fixed-voltage options, connect this pin directly to the output. For an adjustable voltage option, connect this pin to the resistor divider.

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter	Min.	Max.	Unit
$V_{IN}$	Supply Voltage	-0.3	6.5	V
	Input Voltage on PVIN and Any Other Pin	GND-0.2	$V_{IN}+0.3$	V
$\theta_{JC}$	Thermal Resistance <sup>(1)</sup>	Junction-to-Case, SOT-23	130	°C/W
		Junction-to-Tab, MLP 3x3	8	
$T_L$	Lead Soldering Temperature, 10 Seconds		+260	°C
$T_{STG}$	Storage Temperature	-65	+150	°C
ESD <sup>(2)</sup>	Human Body Model, JESD22-A114	4		kV
	Charged Device Model, JESD22-C101	1		

### Notes:

- Junction-to-ambient thermal resistance,  $\theta_{JA}$ , is a strong function of PCB material, board thickness, thickness and number of copper planes, number of via used, diameter of via used, available copper surface, and attached heat sink characteristics.
- Using Mil Std. 883E, method 3015.7 (Human Body Model) and EIA/JESD22C101-A (Charged Device Model).

## Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Min.	Typ.	Max.	Unit
$V_{IN}$	Supply Voltage Range	2.5		5.5	V
$V_{OUT}$	Output Voltage Range, Adjustable Version	0.7		$0.8V_{IN}$	V
$I_{OUT}$	Output Current			300	mA
L	Inductor <sup>(3)</sup>		10		μH
$C_{IN}$	Input Capacitor <sup>(3)</sup>		4.7		μF
$C_{OUT}$	Output Capacitor <sup>(3)</sup>		10		μF
$T_A$	Operating Ambient Temperature Range	-40		+85	°C
$T_J$	Operating Junction Temperature Range	-40		+125	°C

### Note:

- Refer to the *Applications* section for details.

## Electrical Characteristics

$V_{IN}=2.5V$  to  $5.5V$ ,  $I_{OUT}=200mA$ ,  $EN=V_{IN}$ ,  $C_{IN}=4.7\mu F$ ,  $C_{OUT}=22\mu F$ ,  $L_X=10\mu H$ ,  $T_A=-40^{\circ}C$  to  $+85^{\circ}C$ , unless otherwise noted. Typical values are at  $T_A=25^{\circ}C$ .

Symbol	Parameter		Conditions	Min.	Typ.	Max.	Units
$V_{IN}$	Input Voltage			2.5		5.5	V
$I_Q$	Quiescent Current		$I_{OUT}=0mA$ , Device is not switching		15	30	$\mu A$
$I_{SD}$	Shutdown Supply Current		$EN=GND$		0.1	1.0	$\mu A$
$V_{ENH}$	Enable High Input Voltage			1.3			V
$V_{ENL}$	Enable Low Input Voltage					0.5	V
$I_{EN}$	En Input Bias Current		$EN=V_{IN}$ or $GND$		0.01	0.10	$\mu A$
$R_{DS-ON}$	PMOS On Resistance		$V_{IN}=V_{GS}=3.6V$		530	690	$m\Omega$
			$V_{IN}=V_{GS}=2.5V$		670	850	
	NMOS On Resistance		$V_{IN}=V_{GS}=3.6V$		430	540	$m\Omega$
			$V_{IN}=V_{GS}=2.5V$		530	660	
$I_{LIM}$	P-channel Current Limit		$2.5V < V_{IN} < 5.5V$	400	520	700	mA
$I_{IKG(N)}$	N-channel Leakage Current		$V_{DS}=5.5V$		0.1	1.0	$\mu A$
$I_{IKG(P)}$	P-channel Leakage Current		$V_{DS}=5.5V$		0.1	1.0	$\mu A$
	Switching Frequency			800	1000	1200	kHz
$R_{LINE}$	Line Regulation		$V_{IN}=2.5$ to $5.5$ , $I_{OUT}=10mA$		0.16		% / V
$R_{LOAD}$	Load Regulation	6-Lead MLP	$100mA \leq I_{OUT} \leq 300mA$		0.0014		% / mA
		5-Lead SOT-23	$100mA \leq I_{OUT} \leq 300mA$		0.0022		% / mA
$V_{OUT}$	Output Voltage Accuracy	6-Lead MLP	$V_{IN}=2.5$ to $5.5V$ , $0mA \leq I_{OUT} \leq 300mA$	-4		4	%
		5-Lead SOT-23	$V_{IN}=2.5$ to $4.5V$ , $0mA \leq I_{OUT} \leq 300mA$	-4		4	
			$V_{IN}=2.5$ to $5.5V$ , $0mA \leq I_{OUT} \leq 300mA$	-5		4	
$I_{LEAK}$	Leakage Current into Pin SW		$V_{IN} > V_{OUT}$ , $0V \leq V_{SW} \leq V_{IN}$		0.1	1.0	$\mu A$
$I_{LEAK_R}$	Reverse Leakage Current into Pin SW		$V_{IN}=Open$ , $EN=GND$ , $V_{SW}=5.5$		0.1	1.0	$\mu A$

## Electrical Characteristics for Adjustable Version

$V_{IN}=2.5V$  to  $5.5V$ ,  $I_{OUT} = 200mA$ ,  $EN=V_{IN}$ ,  $C_{IN}=4.7\mu F$ ,  $C_{OUT}=22\mu F$ ,  $L_X=10\mu H$ ,  $T_A=25^{\circ}C$ .

Symbol	Parameter	Min.	Typ.	Max.	Units
$V_{FB}$	Feedback Voltage		0.5		V

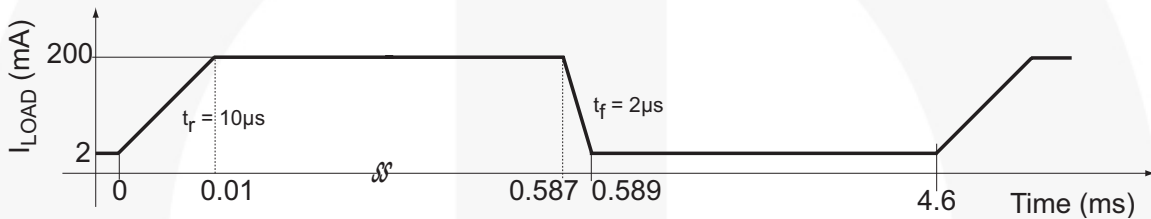
### Electrical Characteristics for Fixed $V_{OUT}=1.8$ Version

$V_{IN}=2.5V$  to  $5.5V$ ,  $I_{OUT}=200mA$ ,  $EN=V_{IN}$ ,  $C_{IN}=4.7\mu F$ ,  $C_{OUT}=22\mu F$ ,  $L_X=10\mu H$ ,  $T_A=-40^\circ C$  to  $+85^\circ C$ , unless otherwise noted. Typical values are at  $T_A=25^\circ C$ .

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$V_{PFM\_PWM}$	PFM to PWM Transition Voltage <sup>(4)</sup>	$V_{IN}=3.7V$ , $T_A=25^\circ C$ , $0.1mA \leq I_{OUT} \leq 300mA$			72	mV
		$V_{IN}=4.2V$ , $T_A=25^\circ C$ , $0.1mA \leq I_{OUT} \leq 300mA$				
$V_{OUT\_TRANS}$	Output Voltage During Mode Transition <sup>(5,6)</sup>		1.70		1.93	V
$V_{OUT\_CLAMP}$	Over-Voltage Clamp Threshold	Includes Line, Load, Load Transients, and Temperature		1.878	1.930	V

**Note:**

4. Transition voltage is defined as the difference between the output voltage measured at 0.1mA (PFM mode) and 300mA (PWM mode), respectively.
5. See Figure 6.
6. These limits also apply to any mode transition caused by any kind of load transition within specified output current range.



**Figure 6. Load Transient Response Test Waveform**

## Typical Performance Characteristics

$T_A=25^\circ\text{C}$ ,  $C_{IN}=C_{OUT}=10\mu\text{F}$ ,  $L=10\mu\text{H}$ ,  $V_{OUT}=1.8\text{V}$ , unless otherwise noted.

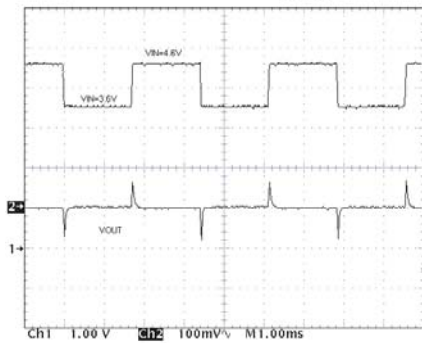


Figure 7. Line Transient Response

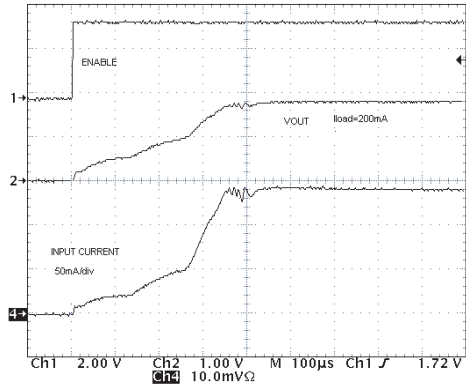


Figure 8. Startup

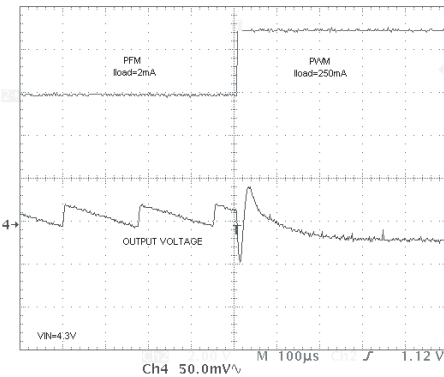


Figure 9. Load Transient Response

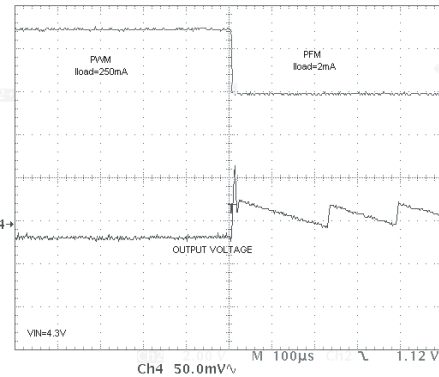


Figure 10. Load Transient Response

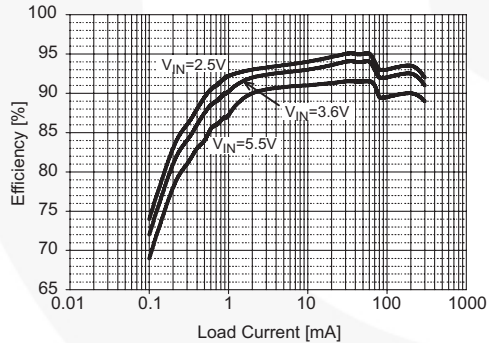


Figure 11. Efficiency vs. Load Current ( $V_{OUT}=1.8\text{V}$ )

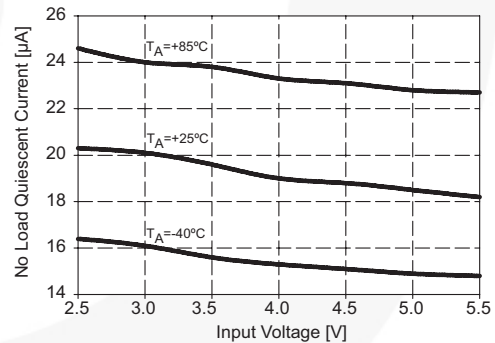


Figure 12. No-Load Quiescent Current vs.  $V_{IN}$

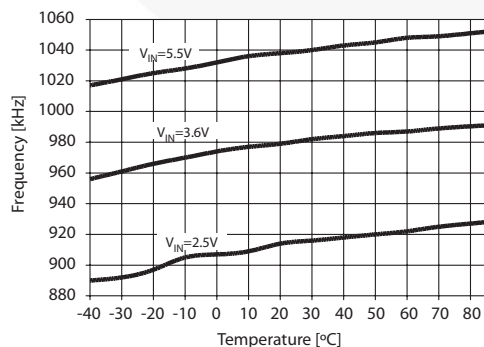


Figure 13. Frequency vs. Temperature

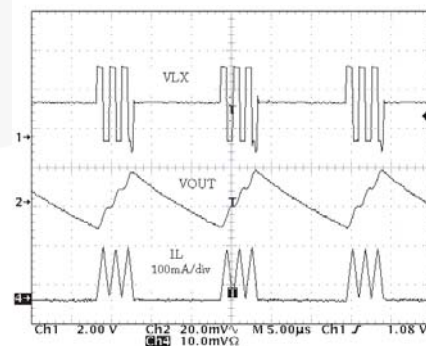


Figure 14. Power Save (PRM) Mode Operation

Block Diagram

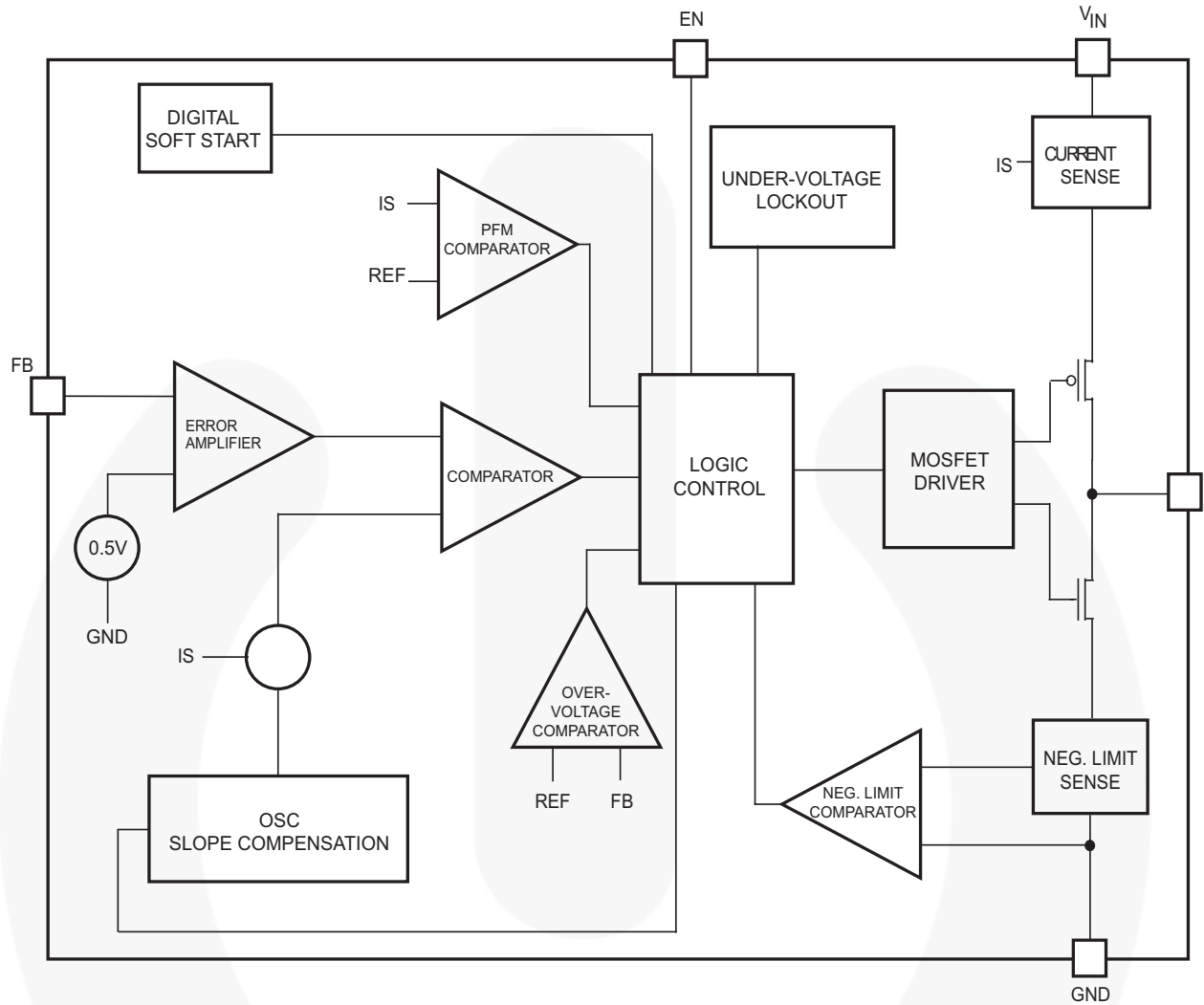


Figure 15. Block Diagram



## Detailed Operation Description

The FAN5307 is a step-down converter operating in a current-mode PFM/PWM architecture with a typical switching frequency of 1MHz. At moderate to heavy loads, the converter operates in pulse-width-modulation (PWM) mode. At light loads, the converter enters a power-save mode (PFM pulse skipping) to keep the efficiency high.

### PWM Mode

In PWM mode, the device operates at a fixed frequency of 1MHz. At the beginning of each clock cycle, the P-channel transistor is turned on. The inductor current ramps up and is monitored via an internal circuit. The P-channel switch is turned off when the sensed current causes the PWM comparator to trip when the output voltage is in regulation or when the inductor current reaches the current limit (set internally to typically 520mA). After a minimum dead time, the N-channel transistor is turned on and the inductor current ramps down. As the clock cycle is completed, the N-channel switch is turned off and the next clock cycle starts.

### FM (Power-Save) Mode

As the load current decreases and the peak inductor current no longer reaches the typical threshold of 80mA, the converter enters pulse-frequency-modulation (PFM) mode. In PFM mode, the device operates with a variable frequency and constant peak current, reducing the quiescent current to a minimum and maintaining high efficiency at light loads. As soon as the output voltage falls below a threshold, set at 0.8% above the nominal value, the P-channel transistor is turned on and the inductor current ramps up. The P-channel switch turns off and the N-channel turns on as the peak inductor current is reached (typical 140mA).

The N-channel transistor is turned off before the inductor current becomes negative. At this time, the P-channel is switched on again, starting the next pulse. The converter continues these pulses until the high threshold is reached (typically 1.6% above nominal value). A higher output voltage in PFM mode gives additional headroom for the voltage drop during a load transient from light to full load. The voltage overshoot during this load transient is minimized due to active regulation during turning on the N-channel rectifier switch. The device stays in sleep mode until the output voltage falls below the low threshold. FAN5307 enters PWM mode as soon as the output voltage can no longer be regulated in PFM with constant peak current.

### 100% Duty Cycle Operation

As the input voltage approaches the output voltage and the duty cycle exceeds the typical 90%, the converter turns the P-channel transistor continuously on. In this mode, the output voltage is equal to the input voltage minus, the voltage drop across the P-channel transistor:

$$V_{OUT} = V_{IN} - I_{LOAD} \times (R_{DSON} + R_L), \text{ where} \quad (1)$$

$R_{DSON}$  = P-channel switch on resistance

$I_{LOAD}$  = Output current

$R_L$  = Inductor DC resistance

### Soft Start

The FAN5307 has an internal soft-start circuit that limits the inrush current during start-up. This prevents possible voltage drops of the input voltage and eliminates the output voltage overshoot. The soft-start is implemented as a digital circuit, increasing the switch current in four steps to the P-channel current limit (520mA). Typical start-up time for a 10 $\mu$ F output capacitor and a load current of 200mA is 500 $\mu$ s.

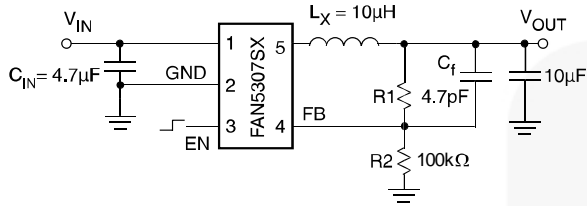
### Short-Circuit Protection

Switch peak current is limited, cycle by cycle, to a typical value of 520mA. In an output voltage short circuit, the device operates at minimum duty cycle; therefore, the average input current is typically 100mA.

## Application Information

### Adjustable Output Voltage Version

The output voltage for the adjustable version is set by the external resistor divider, as shown below:



**Figure 16. External Resistor Divider**

calculated as:

$$V_{OUT} = 0.5V \times \left[ 1 + \frac{R_1}{R_2} \right] \quad (2)$$

To reduce noise sensitivity,  $R_1 + R_2$  should not exceed  $1M\Omega$ .

### Inductor Selection

The inductor parameters directly related to device performance are saturation current and DC resistance. The FAN5307 operates with a typical inductor value of  $10\mu H$ . The lower the DC resistance, the higher the efficiency. For saturation current, the inductor should be rated higher than the maximum load current, plus half of the inductor ripple current, calculated as:

$$\Delta I_L = V_{OUT} \times \frac{1 - (V_{OUT} / V_{IN})}{L \times f} \quad (3)$$

where:

$f$  = Switching frequency

$L$  = Inductor value

$\Delta I_L$  = Inductor ripple current

**Table 1. Recommended Inductors**

Inductor Value	Vendor	Part Number	Performance
10µH	Sumida	CDRH5D28-100	Highest Efficiency
		CDRH5D18-100	
		CDRH4D28-100	
	Murata	LQH66SN100M01L	
6.8µH		CDRH3D16-6R8	Smallest Solution
10µH	Sumida	CDRH4D18-100	
		CR32-100	
		CR43-100	
	Murata	LQH4C100K04	
Cooper Bussmann	CTX01-17327		

### Input Capacitor Selection

For best performances, a low-ESR input capacitor is required. A ceramic capacitor of at least  $4.7\mu F$ , placed as close to the input pin of the device, is recommended.

### Output Capacitor Selection

The FAN5307's switching frequency of 1MHz allows the use of a low-ESR ceramic capacitor with a value of  $10\mu F$  to  $22\mu F$ . This provides low output voltage ripple. In power-save mode, the output voltage ripple is independent of the output capacitor value and the ripple is determined by the internal comparator thresholds. The typical output voltage ripple at light load is 1% of the nominal output voltage.

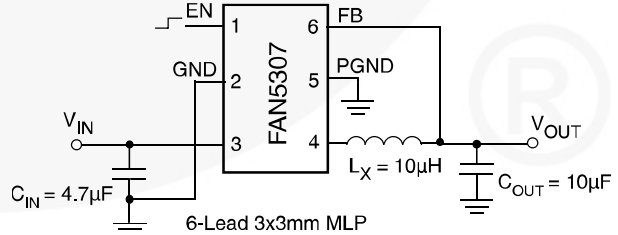
**Table 2. Recommended Capacitors**

Capacitor Value	Vendor	Part Number
4.7µF	Taiyo Yuden	JMK212BY475MG
10µF		JMK212BJ106MG
		JMK316BJ106KL
22µF	TDK	C12012X5ROJ106K
		C3216X5ROJ106M
	Murata	GRM32DR60J226K

### PCB Layout Recommendations

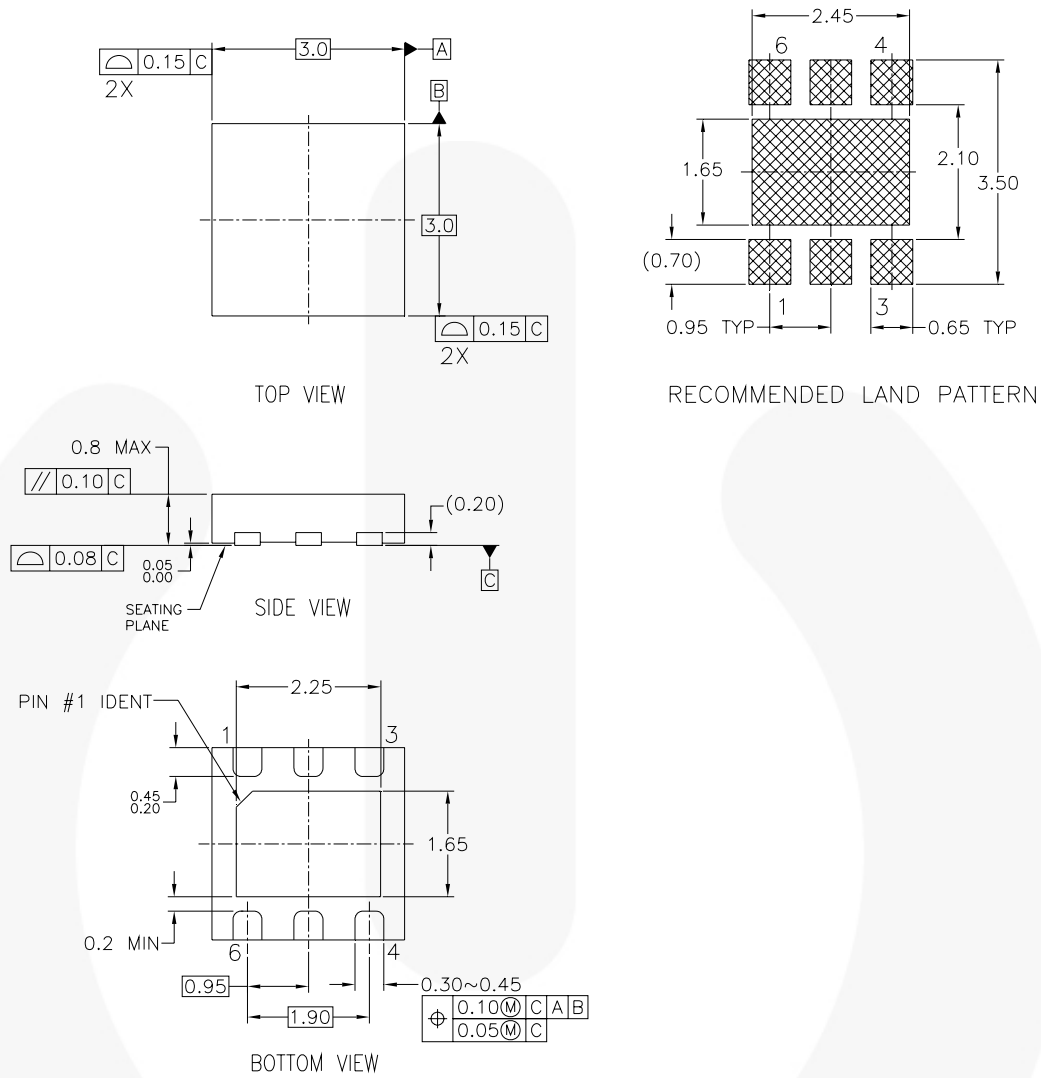
The inherently high peak currents and switching frequency of the power supplies require careful PCB layout design. Use wide traces for the high-current path and place the input capacitor, the inductor, and the output capacitor as close as possible to the integrated circuit terminals. For the adjustable version, the resistor divider should be routed away from the inductor to avoid electromagnetic interference.

The 6-lead MLP version of the FAN5307 separates the high-current ground from the reference ground; therefore, it is more tolerant to the PCB layout design and shows better performance.



**Figure 17. Possible Layout**

Physical Dimensions



NOTES:

- A. CONFORMS TO JEDEC REGISTRATION MO-229, VARIATION WEEA, DATED 11/2001
- B. DIMENSIONS ARE IN MILLIMETERS.
- C. DIMENSIONS AND TOLERANCES PER ASME Y14.5M, 1994

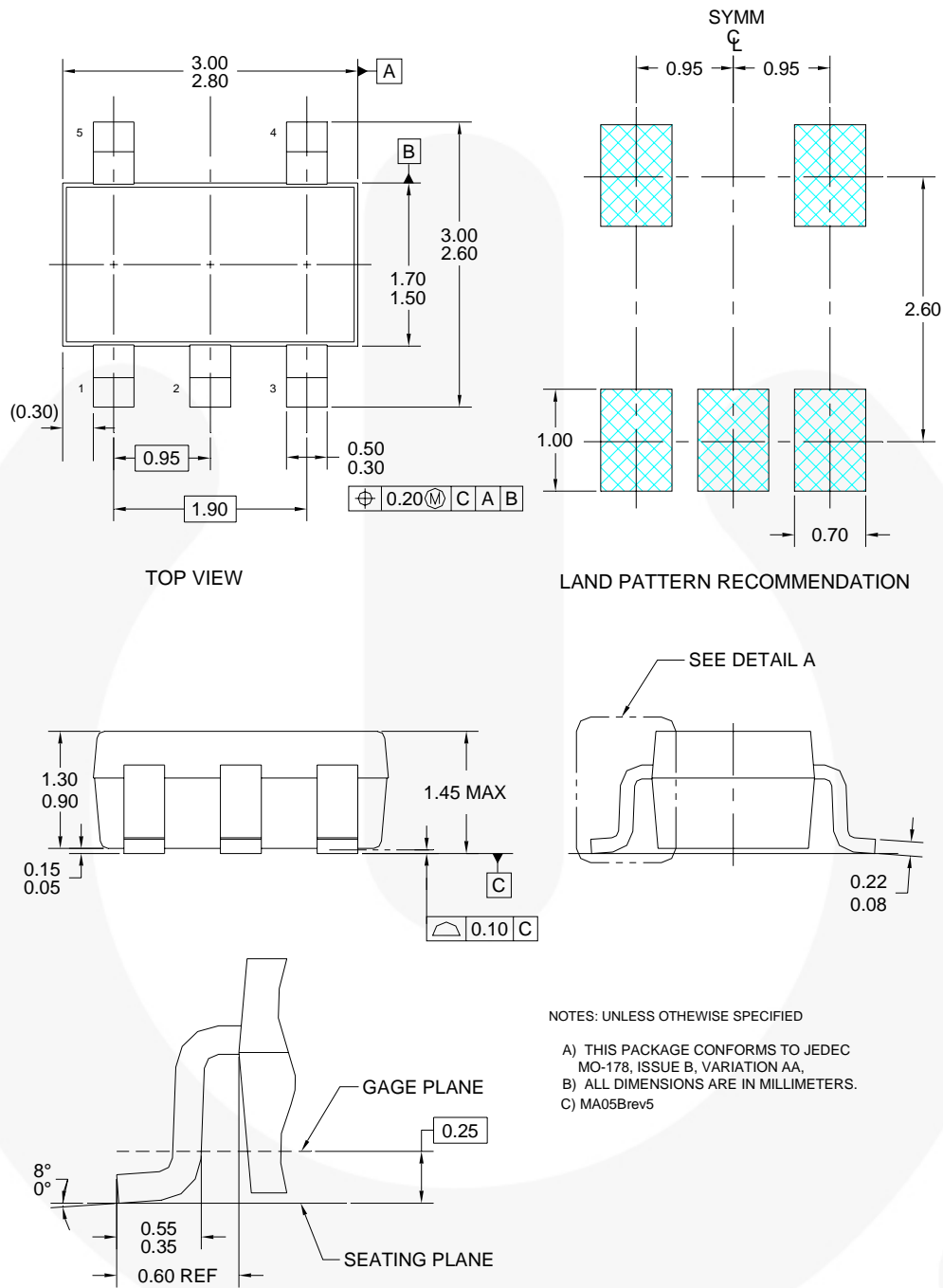
MLP06DrevA

Figure 18. 3x3mm 6-Lead Molded Leadless Package (MLP)

Package drawings are provided as a service to customers considering Fairchild components. Drawings may change in any manner without notice. Please note the revision and/or date on the drawing and contact a Fairchild Semiconductor representative to verify or obtain the most recent revision. Package specifications do not expand the terms of Fairchild's worldwide terms and conditions, specifically the warranty therein, which covers Fairchild products.

Always visit Fairchild Semiconductor's online packaging area for the most recent package drawings:  
<http://www.fairchildsemi.com/packaging/>

### Physical Dimensions



**Figure 19. 5-Lead SOT-23 Package**








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| Build it Now™   | FRFET®  | Programmable Active Droop™  | TinyBoost™  |
| CorePLUS™   | Global Power Resource <sup>SM</sup>   | QFET®   | TinyBuck™   |
| CorePOWER™  | Green FPST™   | QST™  | TinyCalc™   |
| CROSSVOLT™  | Green FPST™ e-Series™   | Quiet Series™   | TinyLogic®  |
| CTL™  | Gmax™   | RapidConfigure™   | TINYOPTO™   |
| Current Transfer Logic™   | GTQ™  |  | TinyPower™  |
| EcoSPARK®   | IntelliMAX™   | Saving our world, 1mW/W at a time™  | TinyPWM™  |
| EfficientMax™   | ISOPLANAR™  | SignalWise™   | TinyWire™   |
| EZSWITCH™   | MegaBuck™   | SmartMax™   | TriFault Detect™  |
|  | MICROCOUPLER™   | SMART START™  | TRUECURRENT™  |
|  | MicroFET™   | SPM®  | μSerDes™  |
| Fairchild®  | MicroPak™   | STEALTH™  |  |
| Fairchild Semiconductor®  | MillerDrive™  | SuperFET™   | UHC®  |
| FACT Quiet Series™  | MotionMax™  | SuperSOT™.3   | Ultra FRFET™  |
| FACT®   | Motion-SPM™   | SuperSOT™.6   | UniFET™   |
| FAST®   | OPTOLOGIC®  | SuperSOT™.8   | VCX™  |
| FastvCore™  | OPTOPLANAR®   | SupreMOS™   | VisualMax™  |
| FETBench™   |  | SyncFET™  | XST™  |
| FlashWriter®  | PDP SPM™  | Sync-Lock™  |   |
|   | Power-SPM™  |  |   |

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**PRODUCT STATUS DEFINITIONS**

**Definition of Terms**

Datasheet Identification	Product Status	Definition
Advance Information	Formative / In Design	Datasheet contains the design specifications for product development. Specifications may change in any manner without notice.
Preliminary	First Production	Datasheet contains preliminary data; supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve design.
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Rev. 142