FAIRCHILD SEMICONDUCTOR®

April 2009

FAN7602C
Green Current Mode PWM Controller

Features

- Green Current Mode PWM Controller
- Random Frequency Fluctuation for Low EMI
- Internal High-Voltage Startup Switch
- Burst Mode Operation
- Line Voltage Feedforward to Limit Maximum Power
- Line Under-Voltage Protection
- Latch Protection & Internal Soft-Start (10ms) Function
- Overload Protection
- Over-Voltage Protection
- Over-Temperature Protection
- Low Operation Current: 1mA Typical
- 8-Pin DIP/SOP

Applications

- Adapter
- LCD Monitor Power
- Auxiliary Power Supply

Related Resources

- *AN-6014- Green Current Mode PWM Controller (Except for frequency fluctuation part in AN-6014)*

Description

The FAN7602C is a green current-mode PWM controller. It is specially designed for off-line adapter applications; DVDP, VCR, LCD monitor applications; and auxiliary power supplies.

The internal high-voltage startup switch and the burst mode operation reduce the power loss in standby mode. As a result, the input power is lower than 1W when the input line voltage is $265V_{AC}$ and the load is 0.5W. At no-load condition, input power is under 0.15W.

The maximum power can be limited constantly, regardless of the line voltage change, using the power limit function.

The switching frequency is not fixed and it has random frequency fluctuation.

The FAN7602C includes various protections for the system reliability and the internal soft-start prevents the output voltage over-shoot at startup.

Ordering Information

Part Number	Operating Junction Temperature	Eco Status	Package	Packing Method	Top Mark
FAN7602CN	-40°C to +150°C	RoHS	8-DIP	Rail	FAN7602C
FAN7602CM			8-SOP	Rail	FAN7602C
FAN7602CMX				Tape and Reel	FAN7602C

For Fairchild's definition of Eco Status, please visit: http://www.fairchildsemi.com/company/green/rohs_green.html.

Typical Application Diagram

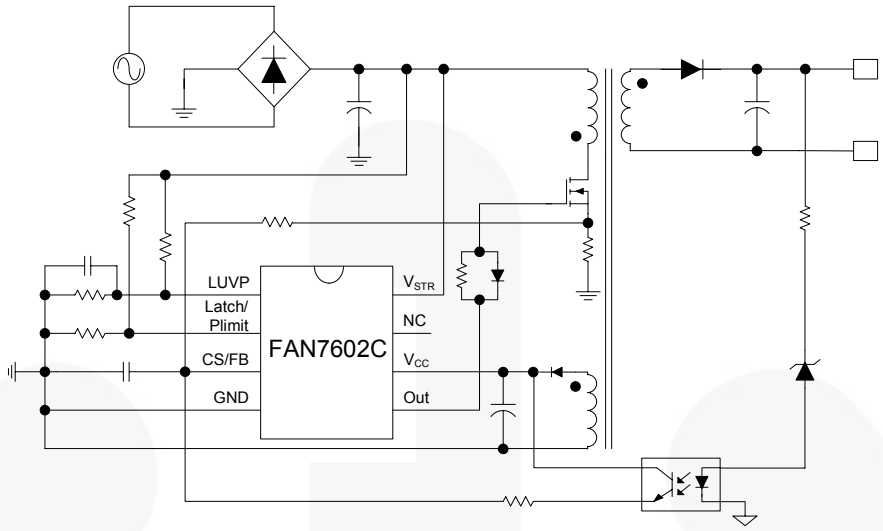


Figure 1. Typical Flyback Application

Internal Block Diagram

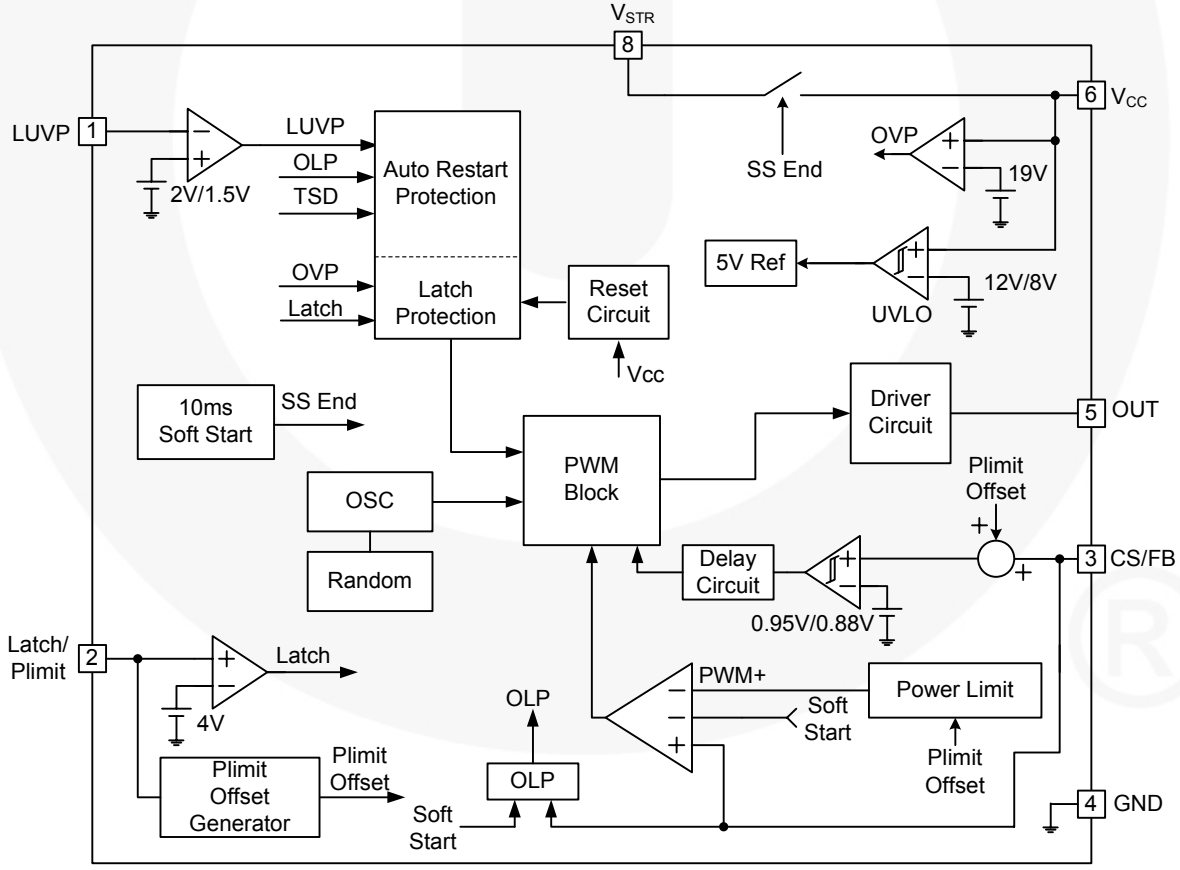


Figure 2. Functional Block Diagram

Pin Configuration

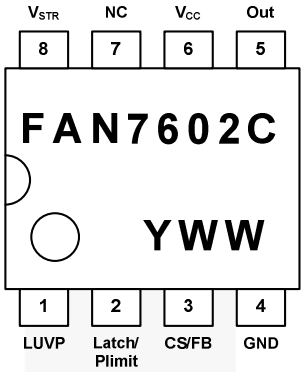


Figure 3. Pin Configuration (Top View)

Pin Definitions

Pin #	Name	Description
1	LUVP	**Line Under-Voltage Protection Pin**. This pin is used to protect the set when the input voltage is lower than the rated input voltage range.
2	Latch/Plimit	**Latch Protection and Power Limit Pin**. When the pin voltage exceeds 4V, the latch protection works. The latch protection is reset when the V_{CC} voltage is lower than 5V. For the power limit function, the OCP level decreases as the pin voltage increases.
3	CS/FB	**Current Sense and Feedback Pin**. This pin is used to sense the MOSFET current for the current mode PWM and OCP. The output voltage feedback information and the current sense information are added using an external RC filter.
4	GND	**Ground Pin**. This pin is used for the ground potential of all the pins. For proper operation, the signal ground and the power ground should be separated.
5	OUT	**Gate Drive Output Pin**. This pin is an output pin to drive an external MOSFET. The peak sourcing current is 450mA and the peak sinking current is 600mA. For proper operation, the stray inductance in the gate driving path must be minimized.
6	V_{CC}	**Supply Voltage Pin**. IC operating current and MOSFET driving current are supplied using this pin.
7	NC	**No Connection**.
8	V_{STR}	**Startup Pin**. This pin is used to supply IC operating current during IC startup. After startup, the internal JFET is turned off to reduce power loss.

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V_{CC}	Supply Voltage			25	V
I_O	Output Current		-600	+450	mA
$V_{CS/FB}$	CS/FB Input Voltage		-0.3	20	V
V_{LUVP}	LUVP Input Voltage		-0.3	10	V
V_{Latch}	Latch/Plimit Input Voltage		-0.3	10	V
V_{STR}	V_{STR} Input Voltage			600	V
T_J	Junction Temperature			+150	°C
	Recommended Operating Junction Temperature		-40	+150	
T_{STG}	Storage Temperature Range		-55	+150	°C
P_D	Power Dissipation			1.2	W
ESD	Electrostatic Discharge Capability	Human Body Model, JESD22-A114		3.5	kV
		Charged Device Model, JESD22-C101		2.0	

Thermal Impedance

Symbol	Parameter		Value	Unit
θ_{JA}	Thermal Resistance[1], Junction-to-Ambient	8-DIP	100	°C/W

Note:

1. Regarding the test environment and PCB type, please refer to JESD51-2 and JESD51-10.

Electrical Characteristics

V_{CC} = 14V, T_A = -25°C~125°C, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
Startup Section						
I_{STR}	V_{STR} Startup Current	V_{STR} = 30V, T_A = 25°C	0.7	1.0	1.4	mA
Under Voltage Lock Out Section						
V_{th}(start)	Start Threshold Voltage	V_{CC} Increasing	11	12	13	V
V_{th}(stop)	Stop Threshold Voltage	V_{CC} Decreasing	7	8	9	V
HY(uvlo)	UVLO Hysteresis		3.6	4.0	4.4	V
Supply Current Section						
I_{ST}	Startup Supply Current	T_A = 25°C		250	320	µA
I_{CC}	Operating Supply Current	Output Not Switching		1.0	1.5	mA
Soft-Start Section						
t_{SS}	Soft-Start Time[2]		5	10	15	ms
PWM Section						
f_{OSC}	Operating Frequency	$V_{CS/FB}$ = 0.2V, T_A = 25°C	59	65	73	kHz
$\triangle f_{OSC}$	Frequency Fluctuation[2]			±3		kHz
$V_{CS/FB1}$	CS/FB Threshold Voltage	T_A = 25°C	0.9	1.0	1.1	V
t_D	Propagation Delay to Output[2]			100	150	ns
D_{MAX}	Maximum Duty Cycle		70	75	80	%
D_{MIN}	Minimum Duty Cycle				0	%
Burst Mode Section						
$V_{CS/FB2}$	Burst On Threshold Voltage	T_A = 25°C	0.84	0.95	1.06	V
$V_{CS/FB3}$	Burst Off Threshold Voltage	T_A = 25°C	0.77	0.88	0.99	V
Power Limit Section						
K_{Plimit}	Offset Gain	$V_{Latch/Plimit}$ = 2V, T_A = 25°C	0.12	0.16	0.20	
Output Section						
V_{OH}	Output Voltage High	T_A = 25°C, I_{source} = 100mA	11.5	12.0	14.0	V
V_{OL}	Output Voltage Low	T_A = 25°C, I_{sink} = 100mA		1.0	2.5	V
t_R	Rising Time[2]	T_A = 25°C, C_L = 1nF		45	150	ns
t_F	Falling Time[2]	T_A = 25°C, C_L = 1nF		35	150	ns

Continued on the following page…

Electrical Characteristics (Continued)

V_{CC} = 14V, T_A = -25°C~125°C, unless otherwise specified.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
Protection Section						
V_{LATCH}	Latch Voltage		3.6	4.0	4.4	V
t_{OLP}	Overload Protection Time[2]		20	22	24	ms
t_{OLP_ST}	Overload Protection Time at Startup		30	37	44	ms
V_{OLP}	Overload Protection Level			0	0.1	V
$V_{LUVPoff}$	Line Under-Voltage Protection On to Off	T_A = 25°C	1.9	2.0	2.1	V
V_{LUVPon}	Line Under-Voltage Protection Off to On	T_A = 25°C	1.4	1.5	1.6	V
V_{OVP}	Over-Voltage Protection	T_A = 25°C	18	19	20	V
T_{SD}	Shutdown Temperature[2]			170		°C
HYS				60		°C

Note:

2. These parameters, although guaranteed, are not 100% tested in production.

Typical Performance Characteristics

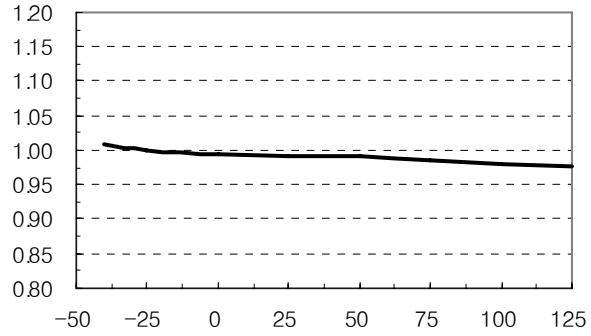


Figure 4. Start Threshold Voltage vs. Temperature

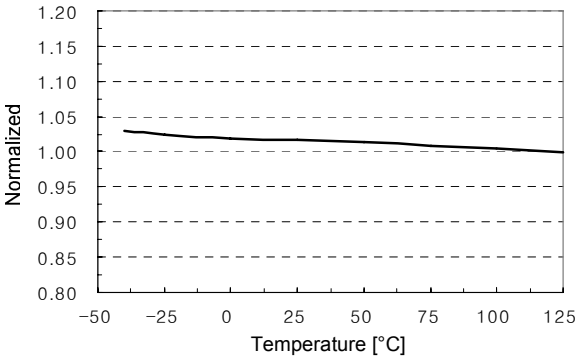


Figure 5. Stop Threshold Voltage vs. Temperature

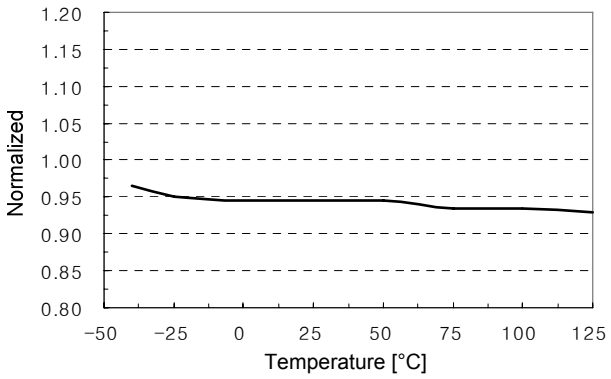


Figure 6. UVLO Hysteresis vs. Temperature

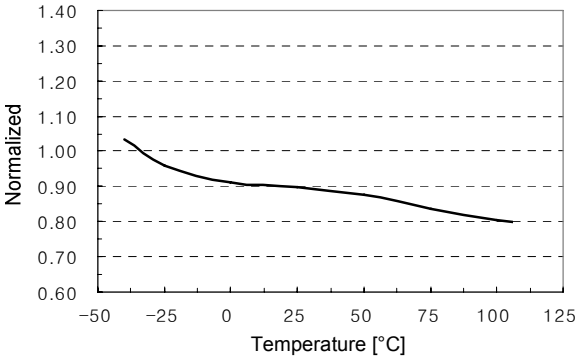


Figure 7. Startup Threshold Current vs. Temperature

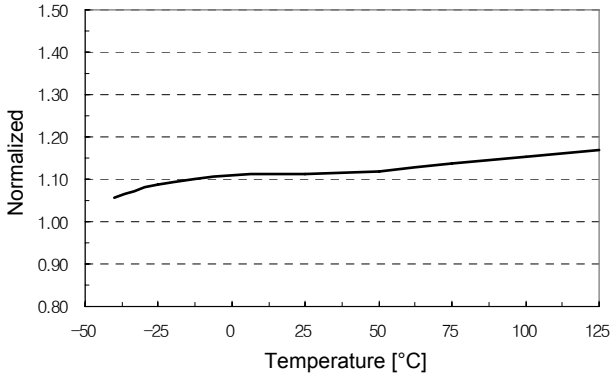


Figure 8. Operating Supply Current vs. Temperature

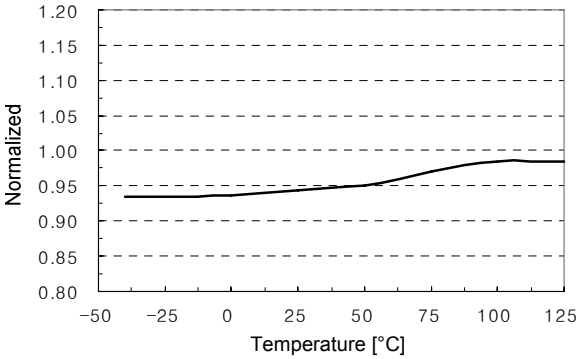


Figure 9. V_{STR} Startup Current vs. Temperature

Typical Performance Characteristics (Continued).

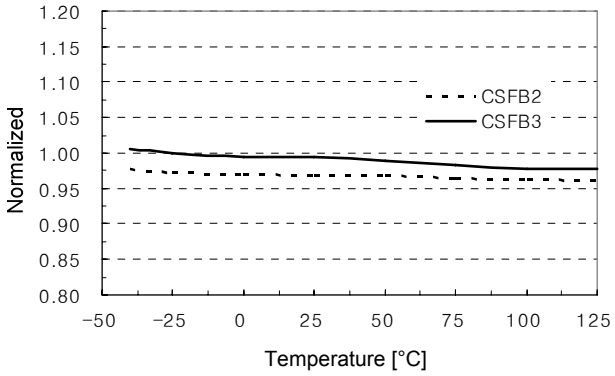


Figure 10. Burst On/Off Voltage vs. Temperature

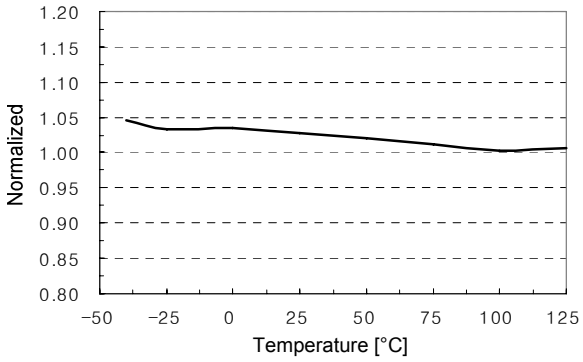


Figure 11. Operating Frequency vs. Temperature

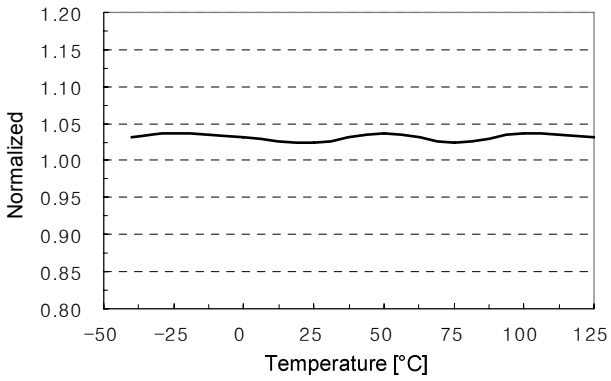


Figure 12. Offset Gain vs. Temperature

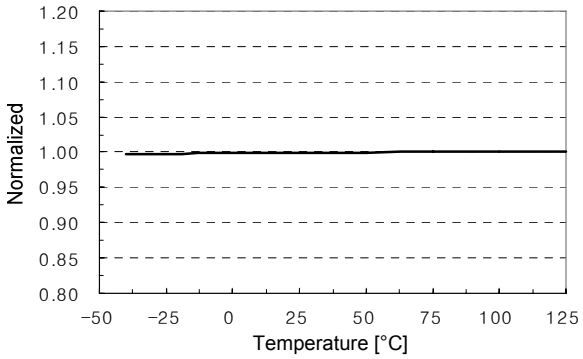


Figure 13. Maximum Duty Cycle vs. Temperature

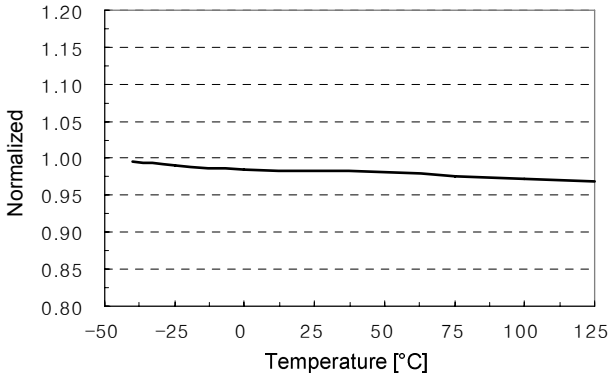


Figure 14. OVP Voltage vs. Temperature

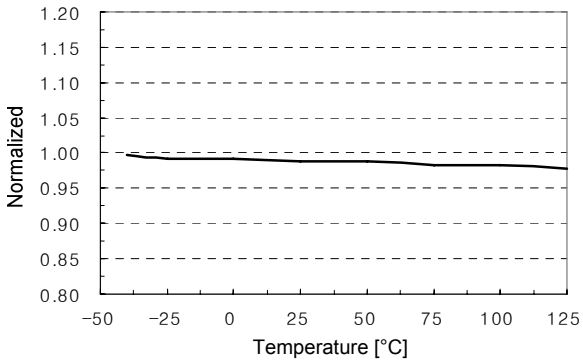


Figure 15. Latch Voltage vs. Temperature

Typical Performance Characteristics (Continued)

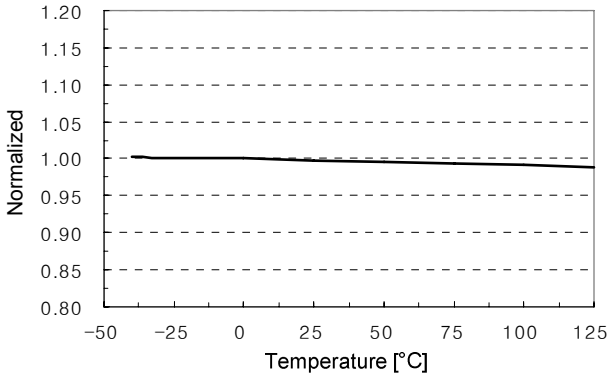


Figure 16. LUVP On-to-Off Voltage vs. Temperature

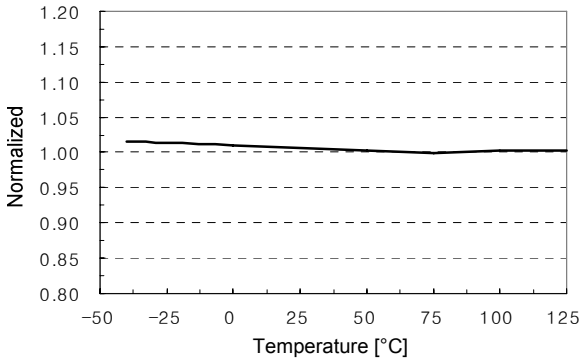


Figure 17. LUVP Off-to-On Voltage vs. Temperature

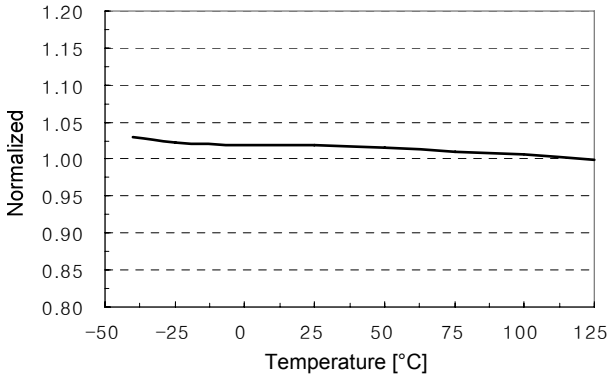


Figure 18. CS/FB Threshold Voltage vs. Temperature

Application Information

1. Startup Circuit and Soft-Start Block

The FAN7602C contains a startup switch to reduce the power loss of the external startup circuit of the conventional PWM converters. The internal startup circuit charges the V_{CC} capacitor with 0.9mA current source if the AC line is connected. The startup switch is turned off 15ms after IC starts up, as shown in Figure 19. The soft-start function starts when the V_{CC} voltage reaches the start threshold voltage of 12V and ends when the internal soft-start voltage reaches 1V. The internal startup circuit starts charging the V_{CC} capacitor again if the V_{CC} voltage is lowered to the minimum operating voltage, 8V. The UVLO block shuts down the output drive circuit and some blocks to reduce the IC operating current and the internal soft-start voltage drops to zero. If the V_{CC} voltage reaches the start threshold voltage, the IC starts switching again and the soft-start block works as well.

During the soft-start, pulse-width modulated (PWM) comparator compares the CS/FB pin voltage with the soft-start voltage. The soft-start voltage starts from 0.5V and the soft-start ends when it reaches 1V and the soft-start time is 10ms. The startup switch is turned off when the soft-start voltage reaches 1.3V.

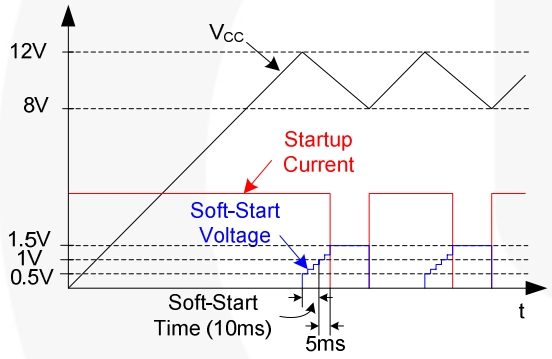


Figure 19. Startup Current and V_{CC} Voltage

2. Oscillator Block

The oscillator frequency is set internally and FAN7602C has a random frequency fluctuation function.

Fluctuation of the switching frequency of a switched power supply can reduce EMI by spreading the energy over a wider frequency range than the bandwidth measured by the EMI test equipment. The amount of EMI reduction is directly related to the range of the frequency variation. The range of frequency variation is fixed internally; however, its selection is randomly chosen by the combination of external feedback voltage and internal free-running oscillator. This randomly chosen switching frequency effectively spreads the EMI noise nearby switching frequency and allows the use of a cost-effective inductor instead of an AC input line filter to satisfy the world-wide EMI requirements.

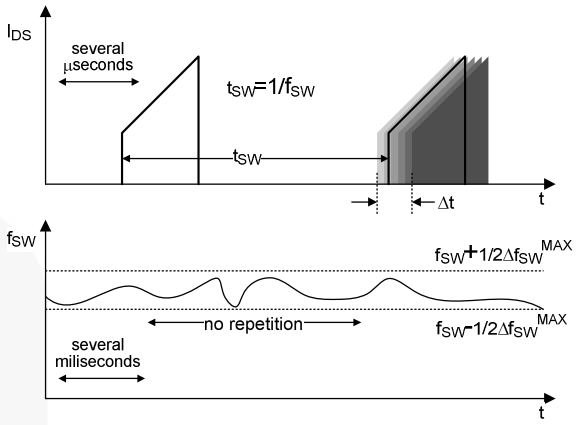


Figure 20. Frequency Fluctuation Waveform

3. Current Sense and Feedback Block

The FAN7602C performs the current sensing for the current mode PWM and the output voltage feedback with only one pin, pin 3. To achieve the two functions with one pin, an internal LEB (leading-edge blanking) circuit to filter the current sense noise is not included because the external RC filter is necessary to add the output voltage feedback information and the current sense information.

Figure 21 shows the current sense and feedback circuits. R_S is the current sense resistor to sense the switch current. The current sense information is filtered by an RC filter composed of R_F and C_F. According to the output voltage feedback information, I_{FB} charges or stops charging C_F to adjust the offset voltage. If I_{FB} is zero, C_F is discharged through R_F and R_S to lower the offset voltage.

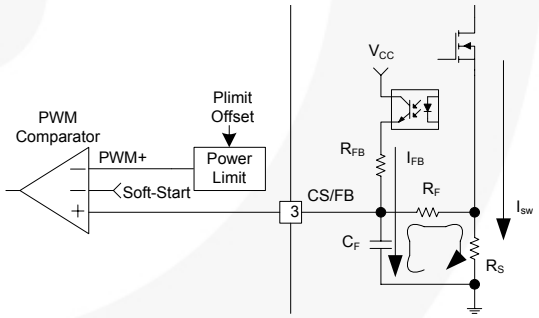
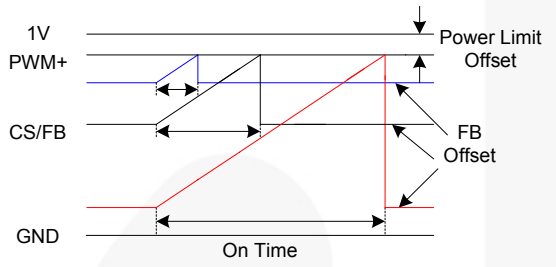


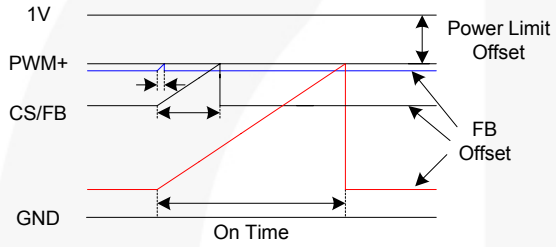
Figure 21. Current Sense and Feedback Circuits

Figure 22 shows typical voltage waveforms of the CS/FB pin. The current sense waveform is added to the offset voltage, as shown in the Figure 22. The CS/FB pin voltage is compared with PWM that is 1V - Plimit offset. If the CS/FB voltage meets PWM+, the output drive is shut off. If the feedback offset voltage is LOW, the switch on-time is increased. If the feedback offset voltage is HIGH, the switch on-time is decreased. In this way, the duty cycle is controlled according to the output load condition. Generally, the maximum output power increases as input voltage increases because the current slope during switch on-time increases.

To limit the output power of the converter constantly, the power limit function is included in FAN7602C. Sensing the converter input voltage through the Latch/Plimit pin, the Plimit offset voltage is subtracted from 1V. As shown in Figure 22, the Plimit offset voltage is subtracted from 1V and the switch on-time decreases as the Plimit offset voltage increases. If the converter input voltage increases, the switch on-time decreases, keeping the output power constant. The offset voltage is proportional to the Latch/Plimit pin voltage and the gain is 0.16. If the Latch/Plimit voltage is 1V, the offset voltage is 0.16V.



(a) Low Power Limit Offset Case



(b) High Power Limit Offset Case

Figure 22. CS/FB Pin Voltage Waveforms

4. Burst Mode Block

The FAN7602C contains the burst-mode block to reduce the power loss at a light load and no load. A hysteresis comparator senses the offset voltage of the Burst+ for the burst mode, as shown in Figure 23. The Burst+ is the sum of the CS/FB voltage and Plimit offset voltage. The FAN7602C enters the burst mode when the offset voltage of the Burst+ is higher than 0.95V and exits the burst mode when the offset voltage is lower than 0.88V. The offset voltage is sensed during the switch off time.

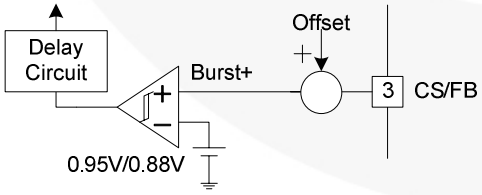


Figure 23. Burst Mode Block

5. Protection Block

The FAN7602C contains several protection functions to improve system reliability.

5.1 Overload Protection (OLP)

The FAN7602C contains the overload protection function. If the output load is higher than the rated output current, the output voltage drops and the feedback error amplifier is saturated. The offset of the CS/FB voltage representing the feedback information is almost zero. As shown in Figure 24, the CS/FB voltage is compared with 50mV reference when the internal clock signal is HIGH and, if the voltage is lower than 50mV, the OLP timer starts counting. If the OLP condition persists for 22ms, the timer generates the OLP signal. The protection is reset by the UVLO. The OLP block is enabled after the soft-start finishes.

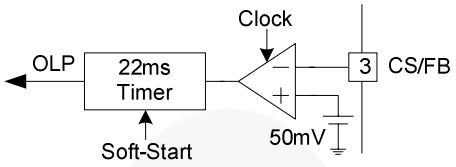


Figure 24. Overload Protection Circuit

5.2 Line Under-Voltage Protection

If the input voltage of the converter is lower than the minimum operating voltage, the converter input current increases too much, causing components failure. Therefore, if the input voltage is LOW, the converter should be protected. The LUVP circuit senses the input voltage using the LUVP pin and, if this voltage is lower than 2V, the LUVP signal is generated. The comparator has 0.5V hysteresis. If the LUVP signal is generated, the output drive block is shut down, the output voltage feedback loop is saturated, and the OLP works if the LUVP condition persists more than 22ms.

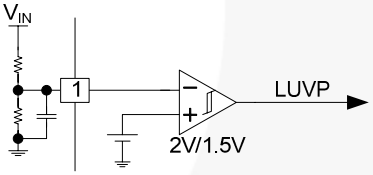


Figure 25. Line UVP Circuit

5.3 Latch Protection

The latch protection is provided to protect the system against abnormal conditions using the Latch/Plimit pin. The Latch/Plimit pin can be used for the output over-voltage protection and/or other protections. If the Latch/Plimit pin voltage is made higher than 4V by an external circuit, the IC is shut down. The latch protection is reset when the V_{CC} voltage is lower than 5V.

5.4 Over-Voltage Protection (OVP)

If the V_{CC} voltage reaches 19V, the IC shuts down and the OVP protection is reset when the V_{CC} voltage is lower than 5V.

6. Output Drive Block

The FAN7602C contains a single totem-pole output stage to drive a power MOSFET. The drive output is capable of up to 450mA sourcing current and 600mA sinking current with typical rise and fall time of 45ns and 35ns, respectively, with a 1nF load.

Typical Application Circuit

Application	Output Power	Input Voltage	Output Voltage
Adaptor	48W	Universal input (85 ~ 265 V_{AC})	12V

Features

- *Low stand-by power (<0.15W at 265 V_{AC})*
- *Constant output power control*

Key Design Notes

- *All the IC-related components should be placed close to IC, especially C107 and C110.*
- *If R106 value is too low, there can be subharmonic oscillation.*
- *R109 should be designed carefully to make the V_{CC} voltage higher than 8V when the input voltage is 265 V_{AC} at no load.*
- *R110 should be designed carefully to make the V_{CC} voltage lower than OVP level when the input voltage is 85 V_{AC} at full load.*
- *R103 should be designed to keep the MOSFET V_{ds} voltage lower than maximum rating when the output is shorted.*

1. Schematic

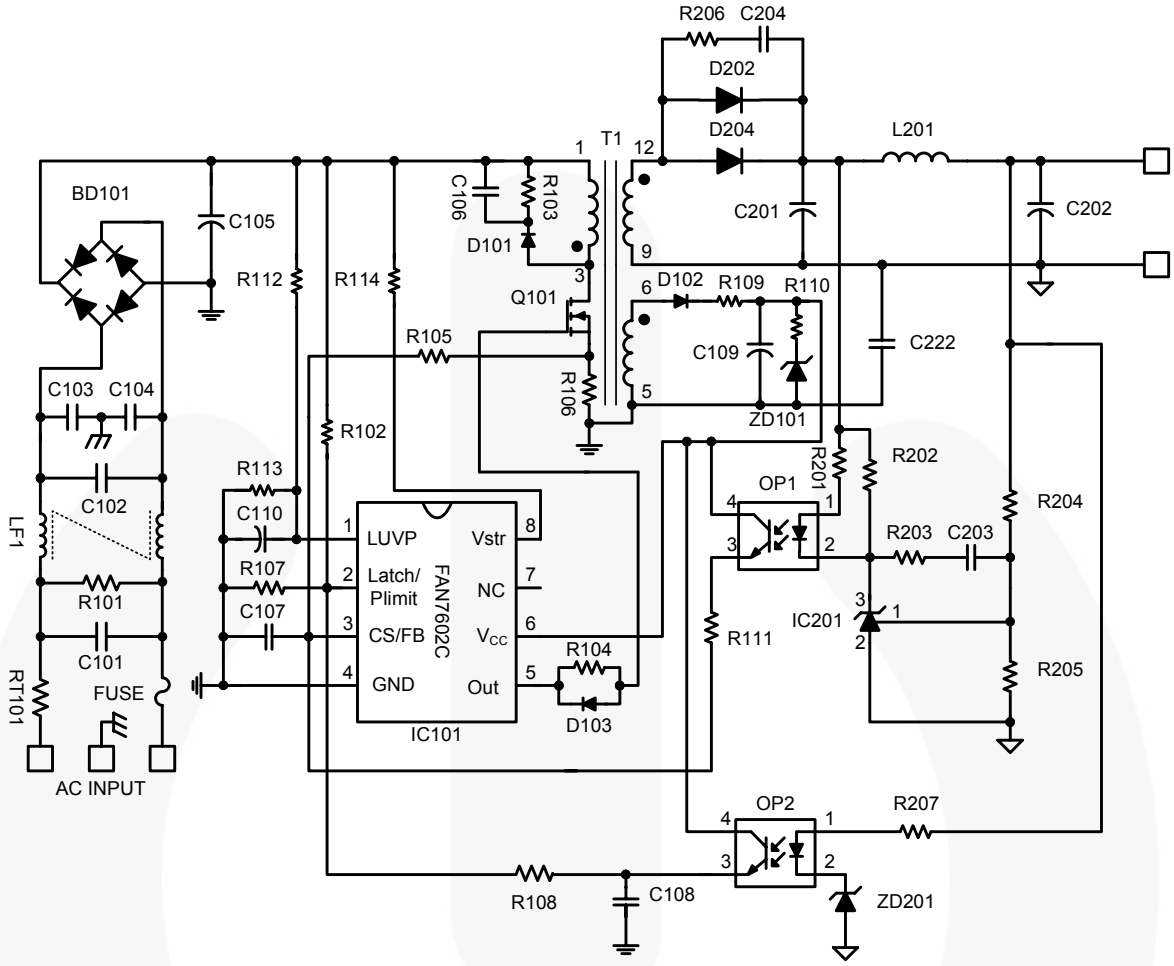


Figure 26. Schematic

2. Inductor Schematic Diagram

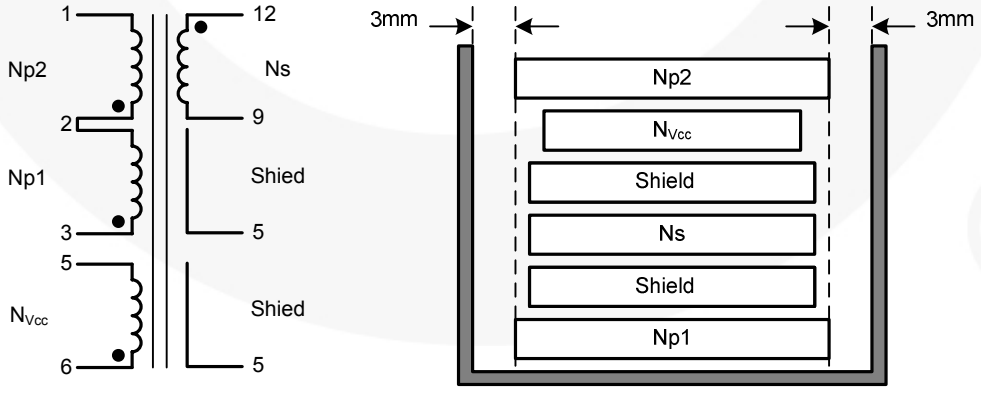


Figure 27. Inductor Schematic Diagram

3. Winding Specification

No	Pin (s → f)	Wire	Turns	Winding Method
N_{p1}	3 → 2	0.3^{ϕ} x 2	31	Solenoid Winding
Insulation: Polyester Tape t = 0.03mm, 2 Layers				
Shield	5	Copper Tape	0.9	Not Shorted
Insulation: Polyester Tape t = 0.03mm, 2 Layers				
N_s	12 → 9	0.65^{ϕ} x 3	10	Solenoid Winding
Insulation: Polyester Tape t = 0.03mm, 2 Layers				
Shield	5	Copper Tape	0.9	Not Shorted
Insulation: Polyester Tape t = 0.03mm, 2 Layers				
N_{Vcc}	6 → 5	0.2^{ϕ} x 1	10	Solenoid Winding
Insulation: Polyester Tape t = 0.03mm, 2 Layers				
N_{p2}	2 → 1	0.3^{ϕ} x 2	31	Solenoid Winding
Outer Insulation: Polyester Tape t = 0.03mm, 2 Layers				

4. Electrical Characteristics

	Pin	Specification	Remarks
Inductance	1 - 3	607μH	100kHz, 1V
Inductance	1 - 3	15μH	9 - 12 shorted

5. Core & Bobbin

- *Core: EER2828*
- *Bobbin: EER2828*
- *Ae(mm^2): 82.1*

6. Demo Circuit Part List

Part	Value	Note	Part	Value	Note
Fuse			**Capacitor**		
FUSE	1 A/250 V		C101	220nF / 275V	Box Capacitor
NTC			C102	150nF / 275V	Box Capacitor
RT101	5D-9		C103, C104	102 / 1kV	Ceramic
Resistor			C105	150μF / 400V	Electrolytic
R102, R112	10MΩ	1/4W	C106	103 / 630V	Film
R103	56kΩ	1/2W	C107	271	Ceramic
R104	150Ω	1/4W	C108	103	Ceramic
R105	1kΩ	1/4W	C109	22μF / 25V	Electrolytic
R106	0.5Ω	1/2W	C110	473	Ceramic
R107	56kΩ	1/4W	C201, C202	1000μF / 25V	Electrolytic
R108	10kΩ	1/4W	C203	102	Ceramic
R109	0Ω	1/4W	C204	102	Ceramic
R110	1kΩ	1/4W	C222	222 / 1kV	Ceramic
R111	6kΩ	1/4W	**MOSFET**		
R113	180kΩ	1/4W	Q101	FQPF8N60C	Fairchild
R114	50kΩ	1/4W	**Diode**		
R201	1.5kΩ	1/4W	D101, D102	UF4007	Fairchild
R202	1.2kΩ	1/4W	D103	1N5819	Fairchild
R203	20kΩ	1/4W	D202, D204	FYPF2010DN	Fairchild
R204	27kΩ	1/4W	ZD101, ZD201	1N4744	Fairchild
R205	7kΩ	1/4W	BD101	KBP06	Fairchild
R206	10Ω	1/2W	**TNR**		
R207	10kΩ	1/4W	R101	471	470V
IC			**Filter**		
IC101	FAN7602C	Fairchild	LF101	23mH	0.8A
IC201	KA431	Fairchild	L201	10μH	4.2A
OP1, OP2	H11A817B	Fairchild			

7. PCB Layout

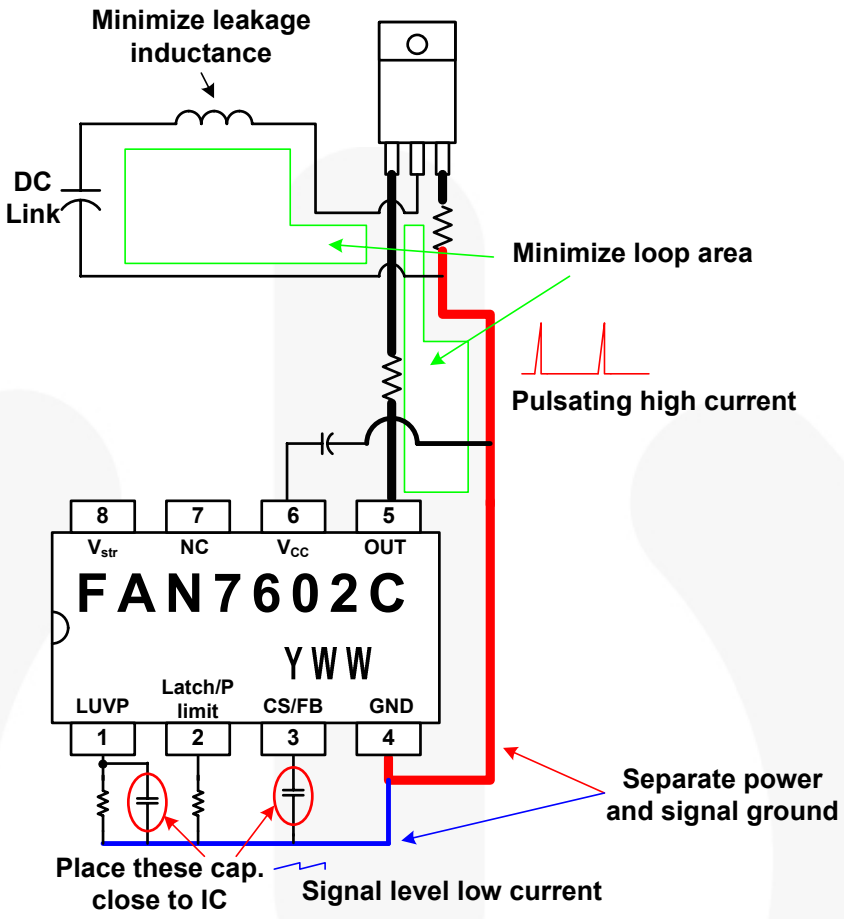
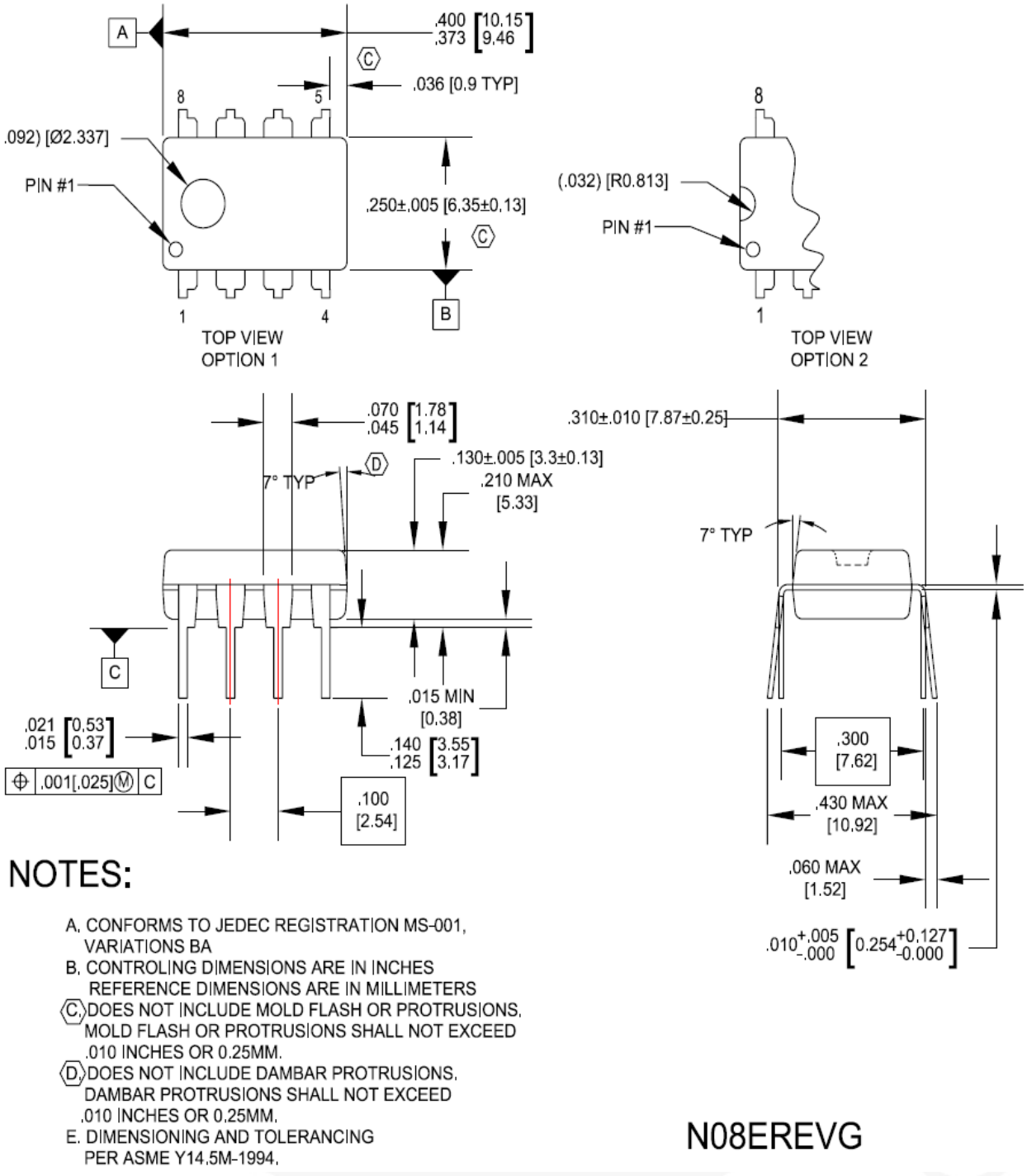


Figure 28. PCB Layout Recommendations

8. Performance Data

	$85V_{AC}$	$110V_{AC}$	$220V_{AC}$	$265V_{AC}$
Input Power at No Load	72mW	76mW	92mW	107mW
Input Power at 0.5W Load	760mW	760mW	785mW	805mW
OLP Point	4.73A	5.07A	5.11A	4.91A

Physical Dimensions



NOTES:

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C. DOES NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCHES OR 0.25MM.
D. DOES NOT INCLUDE DAMBAR PROTRUSIONS. DAMBAR PROTRUSIONS SHALL NOT EXCEED .010 INCHES OR 0.25MM.
E. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.

N08EREVG

Figure 29. 8-Lead Dual In-Line Package (DIP)

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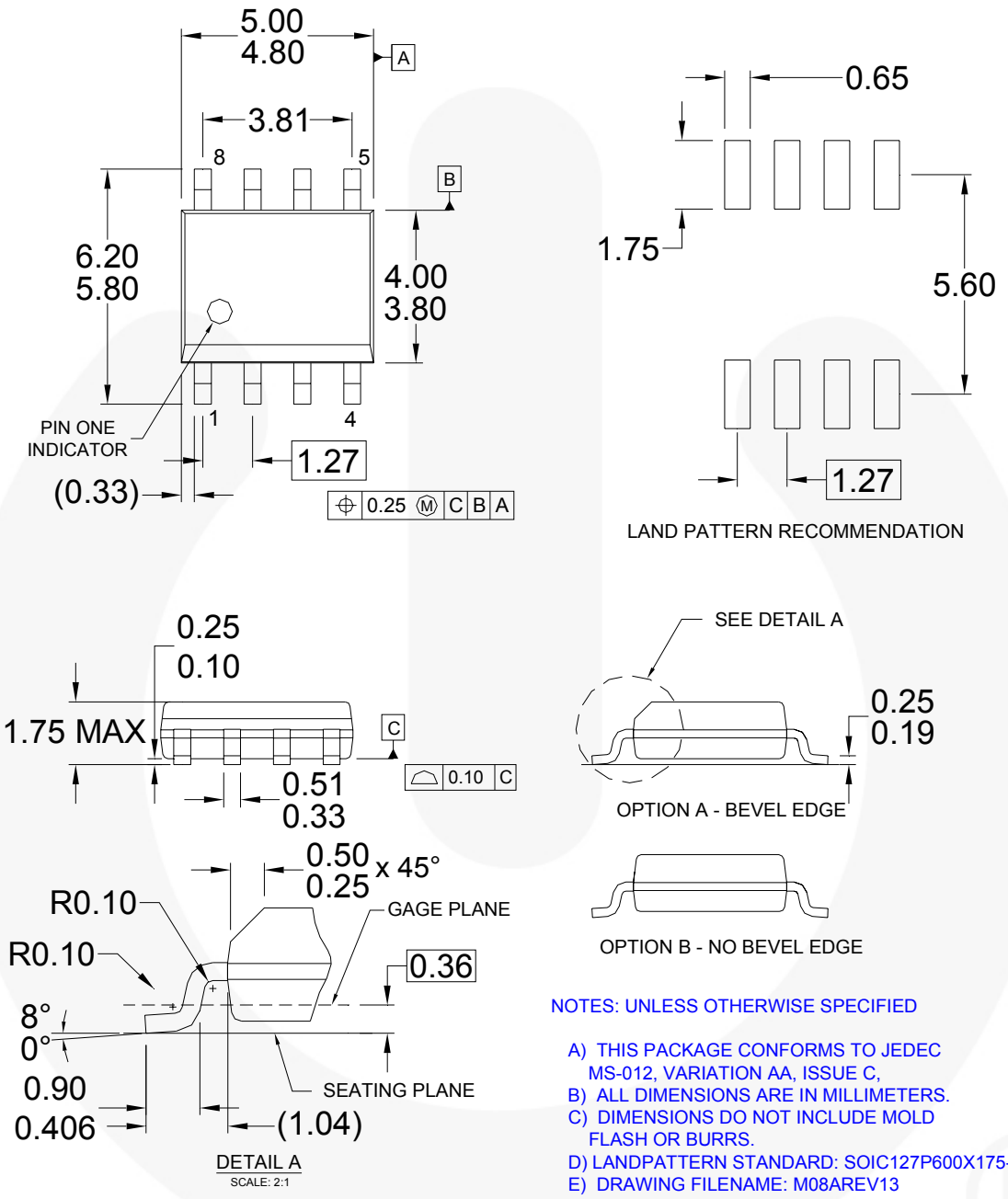
Physical Dimensions (Continued)



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C) DIMENSIONS DO NOT INCLUDE MOLD FLASH OR BURRS.
D) LANDPATTERN STANDARD: SOIC127P600X175-8M.
E) DRAWING FILENAME: M08AREV13

Figure 30. 8-Lead Small Outline Package (SOP)

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